# Mitigation of Power Quality Issues using UPQC with STF-SRF Technique Under Adverse Voltage Source Condition

# Muhammad Alif Mansor, Muhammad Murtadha Othman, Ismail Musirin and Siti Zaliha Mohammad Noor.

Abstract— A three-phase three-wire of Unified Power Quality Conditioner (UPQC) is used to compensate for the power quality issues that exist in the point of common coupling (PCC) on the microgrid (MG) of the power system and the harmonics penetrated by the non-linear loads. This paper presents an amalgamation of a self-tuning filter (STF) with synchronous reference framework (SRF) of direct-quadrature-zero (dq0) fundamental based on UPQC to enhance power quality at the PCC under adverse voltage source condition (unbalanced and distorted source voltage). The modified SRF technique integrated the selftuning filter (STF) with the unit vector generator (UVG) technique for designing the control phase synchronization algorithm of UPQC. Therefore, a phase-locked loop (PLL) necessity is omitted. The STF-UVG is used to produce the synchronization phases reference which can lead to UPQC controller generated reference current and voltage in phase with the operating power system. The suggested STF-SRF technique is compared with the traditional SRF technique by considering several case scenarios of source voltage in MATLAB-Simulink software are discussed in detail.

Index Terms- Power Quality Issues, Unified Power Quality Conditioner (UPQC), Self-Tuning Filter (STF), Synchronization Phases, Phase-Locked Loop (PLL), Microgrid (MG).

## I. INTRODUCTION

THE booming global energy demand is the main reason for I increased fossil fuel burning and increased greenhouse gas emissions. Many researchers are working to find ways to substitute traditional fossil fuels and reduce environmental issues by encouraging the renewable generation sector. For this rationale, in recent years, the implementation and unification of renewable energy generation (REG) into the existing power system has grown significantly. The amalgamation of existing distribution generation (DG) with energy storage systems (ESS) is also needed due to REG characteristics of high environmentdependency, instability, and generate a small amount of power during peak time [1][2]. Therefore, MG's concept is developed with composed of REG, ESS, DG, and load cluster that share the same PCC that performs as a single grid as shown in Fig. 1.

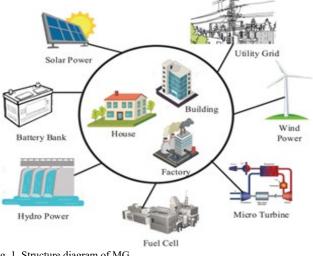


Fig. 1. Structure diagram of MG.

However, one of the most relevant technical challenges with functioning and handling MG systems concerns power quality issues [3]. Due to the structure, operating model, and REG performance in MG, these challenges are a significant concern in the MG system. Many of the power quality issues caused by DG's high penetration include current harmonics, voltage harmonics, voltage swell or sag, fluctuation, unbalance, system breakdown, protective electrical equipment overloading, and failure [4]. This matter is caused by MG sources entirely dependent on power electronic converters, which may also extend the presence of power electronic interfaces in an existing distribution network [5]. One of the most powerful solutions to enhance the power quality is to utilize the UPQC on the PCC of the MG [6]. It is a custom power device that can mitigate the power quality issues on both sides of voltage and current [7]. H. Akagi first introduced the UPQC in 1995 as one of the most innovative technologies that

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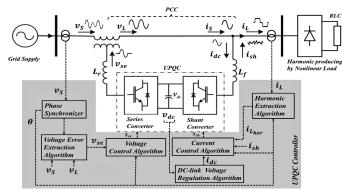


Fig. 2. UPQC overall main control scheme structure

can deal with both supply voltage imperfection on the PCC and load current contaminated by harmonic from a non-linear load [8]. The primary purpose of the UPQC is to minimize the disturbances that affected the operation of critical and sensitive load in the power system. The primary circuit configuration of UPQC with a series and shunt active power filter (APF) compensator with a share typical DC-link capacitor is displayed in Fig. 2. Furthermore, the series APF compensator's primary function is compensation of voltage unbalanced and disturbance and voltage regulation at both sides of consumer and utility. Other than that, it also can maintain harmonic isolation between supply at PCC and load [9][10]. Moreover, the shunt APF compensator is optimized for compensating current harmonics polluted by non-linear load and reactive power. It is also utilized to regulate the DC-link voltage capacitor for the stability of UPQC operation [11][12]. The configuration of UPQC is consist of the power circuit topologies and control scheme structure. In the power circuit topologies of UPQC, there is a voltage source converter (VSC), inductor as passive ripple filter, and isolation transformer as system protection [13]. Meanwhile, the control scheme structure of UPQC consists of six main control algorithms: phase synchronization algorithm, harmonic extraction algorithm, current control algorithm, voltage error extraction algorithm, voltage control algorithm, and DC-link voltage regulation algorithm. The effectiveness of UPQC is all reliant on the ability of its closed-loop control system. There are two techniques that are reported to be extensively applied in the three-phase three-wire system as control-based method namely SRF technique of dq0 fundamental and instantaneous power pq0 theory. Despite that, dq0 fundamental is preferred in this work for its reduced control complexity. To UPQC function effectively, the control system generating reference voltage and current needs to be given priority. Note that generating reference voltage and current always comes together with the process of phases synchronization and extracting harmonic from the operating power system. This is because compensation of voltage and current should be inject in phase correctly with operating power system which is needed phase synchronization algorithm in UPQC controller by identifies a phase synchronization angle and frequency for voltage and current control algorithm especially performing in the dq0-frames transformation. Other than that, harmonic extraction algorithm was for removing the distorted signal to control the operation

of the UPQC in minimizing the harmonic distortion. To determine the qualities of the generated reference voltage and current depend on extracting a fundamental element from the load current and the source voltage for extracting harmonic distortion and phases synchronization. The source voltage and load current comprise fundamental and harmonic elements. A high ripple level tends to degrade the technique's capability and quality of the generated reference current. However, a highpass filter (HPF) commonly applied in traditional SRF-PLL to segregate the fundamental and harmonic elements may not be effective even though a good agreement between the filter's order and its cut-off frequency has been met. Moreover, many issues exist in matching the filter's order and cut-off frequency as it is performed iteratively. In other words, there is a need to test various combinations of the filter's order and cutting frequency to find out the best combination. As a result, a large amount of time will be spent just for the tuning purposes, and it is not worthwhile to be implemented, which causes unwanted ripples in the reference current and voltage. For this phase synchronization algorithm, the PLL technique is generally used and most widely known [14]. Most UPQC controllers are builtin [15]-[19] and [20] using a traditional SRF-PLL for the synchronization phase algorithm, which is cannot performance effectively as desired under unbalance source voltage condition and unable to handle distorted source voltage conditions. The structure controller of traditional SRF-PLL contains a phase detector (PD), a loop filter (LF), and a voltage-controlled oscillator (VCO) shown in Fig. 3. In general, a PI controller as a part of VCO of SRF-PLL will increase the control's complexity and needs fine-tuning, which takes time. An alternative to solve phase tracking error based on UPQC of a traditional PLL is improving its tracking capability under unbalanced and distorted source voltage conditions. These have led researchers to develop a phase synchronization algorithm based on UPQC which is negative-feedback PLL [21], enhanced PLL [22] and sliding Goertzel discrete Fourier transformation-pre-filtering PLL [23] but was increases the complexity of an already complex PLL structure. In the UPQC control scheme, the process algorithm of phase synchronization is one of the most critical aspects that need to consider. By injecting the voltage and current in phase with the grid, UPQC can effectively perform the mitigation process.

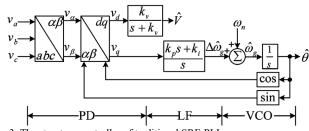


Fig. 3. The structure controller of traditional SRF-PLL

In this study, a potential solution for the phase synchronization algorithm is the STF, which has excellent phase tracking and specific, fundamental component extraction capabilities for dealing with adverse source voltage conditions. To remove the complexity working block VCO of PLL, the straightforward phase synchronization of UVG is utilized to generate unit vectors consisting of sine and cosine functions. Furthermore, the modified traditional SRF technique with STF is utilized to extract harmonic elements to generate reference current for UPQC. The suggested STF-SRF technique based on UPQC is compared with the traditional SRF technique based on UPQC. Validation of this study to support the STF-SRF technique based on UPQC was done by considering several case source voltage scenarios in MATLAB-Simulink software is discussed in detail.

## II. PROPOSED STF-SRF TECHNIQUE BASED ON UPQC

The main controller of UPQC is the series and shunt APF compensator controller, which includes the synchronization phase algorithm method for both controllers. The series APF compensator controller insulates the critical load from the source voltage of power quality problems (grid side) by compensating appropriate voltage in phase with the grid voltage. The shunt APF compensator is optimized for compensating current harmonics polluted by non-linear load and reactive power. It also is utilized for regulating the DC-link voltage capacitor for the stability of UPOC operation. UPOC is a control-based method designed to integrate STF with the SRF technique of dq0-fundamental to perform effectively under unbalanced and distorted source voltage conditions. Generally, the present STF-SRF technique based on UPQC generates reference voltage and current in the same concept as a traditional SRF technique. Nevertheless, the two improvements are made to enhance further the performance of UPQC, which utilizes the straightforward phase synchronization of UVG with STF (STF-UVG) for suppressing the negative sequence of the source voltage. Other than that, STF also utilizes to extract the harmonic element for generating reference current in the shunt APF compensator controller.

## A. Synchronization Phase Algorithm Method

The suggested STF-UVG utilizes the measured source voltage at the PCC to generate the synchronization phase for the STF-SRF technique based on UPQC to compensate in-phase with the grid supply shown in Fig. 4. The Clarke transformation matrix is utilized to transform three-phase source voltage from abc-domain to  $\alpha\beta0$ -domain is demonstrated in (1).

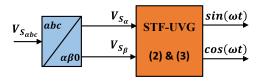


Fig. 4. Synchronization Phase control scheme structure

$$\begin{bmatrix} U_{\alpha} \\ U_{\beta} \\ U_{0} \end{bmatrix} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & -\frac{1}{2} & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \\ \frac{1}{2} & \frac{1}{2} & \frac{1}{2} \end{bmatrix} \begin{bmatrix} U_{a} \\ U_{b} \\ U_{c} \end{bmatrix}$$
(1)

By considering only two phases in the  $\alpha\beta$ -domain, the distorted source voltage can be categories into a fundamental and harmonic element. Therefore, the fundamental (fund) element state as a  $V_{s,\alpha\beta(fund)}$  and the harmonic (har) element state as  $V_{s,\alpha\beta(fund)}$  in  $\alpha\beta$ -domain. It is required the fundamental element only to produce phase reference. By using the STF, the fundamental element can split from the distorted source voltage signal. Thus, the STF can eliminate all the negative sequences existing in the distorted voltage signal and improve the synchronization phase's extraction quality. The Laplace transformation is executed, and a typical STF transfer function is shown in (2)

$$\begin{bmatrix} U_{\alpha \ (fund)}(s) \\ U_{\beta \ (fund)}(s) \end{bmatrix} = \frac{K_1}{s} \begin{bmatrix} U_{\alpha}(s) - U_{\alpha \ (fund)}(s) \\ U_{\beta}(s) - U_{\beta \ (fund)}(s) \end{bmatrix} + \frac{2\pi f_{c1}}{s} \begin{bmatrix} -U_{\beta \ (fund)}(s) \\ U_{\alpha \ (fund)}(s) \end{bmatrix}$$
(2)

Here,  $K_1$  indicates the constant gain parameter, and  $f_{c1}$  indicates the power system's cut-off frequency. In this study, the rating is determined 20 for  $K_1$  and 50 Hz for  $f_{c1}$ . In (3), demonstrated the straightforward phase synchronization of UVG, which applying the  $V_{s,\alpha(fund)}$  and  $V_{s,\beta(fund)}$  to obtain synchronization phases reference in term of  $\sin(\omega t)$  and  $\cos(\omega t)$ . The action of a traditional VCO of PLL structure can be omitted, which can reduce control complexity.

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$$\begin{bmatrix} \sin(\omega t) \\ \cos(\omega t) \end{bmatrix} = \frac{1}{\sqrt{(V_{S_{\alpha}(fund)})^{2} + (V_{S_{\beta}(fund)})^{2}}} \begin{bmatrix} V_{S_{\alpha}(fund)} \\ -V_{S_{\beta}(fund)} \end{bmatrix}$$
(3)

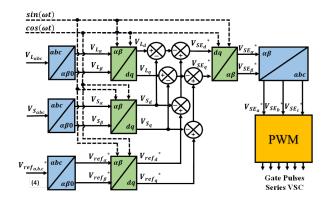


Fig. 5. Series APF Compensator control scheme structure

## B. Series APF Compensator Controller

The control scheme structure of series APF compensator based on SRF technique implement the following sequence of transformation ( $abc-\alpha\beta0-dq0-\alpha\beta0-abc$ ) displayed in Fig. 5. Firstly, in (4), is the three-phase reference voltage signal,  $V_{ref,abc}^*$  in *abc*-domain. Here, the parameter of maximum peak voltage magnitude,  $V_{m,max-peak}$  is acquired from a peak amplitude of the actual grid source voltage. Next, the threephase reference voltage signal converts from abc-domain to  $\alpha\beta0$ -domain using the Clarke transformation matrix (1). Secondly, the resulted synchronization phases reference STF-UVG in term of  $sin(\omega t)$  and  $cos(\omega t)$  obtained in (3) is utilized for creating a reference axis in the dq-domain of the park transformation matrix. It is used to a covert reference voltage signal in the dq-domain id demonstrated in (5). Furthermore, the three-phase load voltage,  $V_{L,\alpha bc}$  and the three-phase supply  $V_{s,\alpha bc}$  also, utilizing the Clarke-Park matrix to voltage, transform from abc-domain into dq-domain is demonstrated in (1) and (5). In the series APF compensator, a voltage error extraction algorithm method is using a comparison technique of the load voltage,  $V_{L,dq}$  and source voltage,  $V_{s,dq}$  at PCC, both in the dq-domain to result in real voltage error. Next, a real reference voltage is obtained by differentiating between the reference voltage signal, V<sub>ref,dq</sub>\* and supply voltage, V<sub>s,dq</sub> in the dq-domain. Thus, the compensation reference voltage,  $V_{SE,da}$ acquired by comparing a real reference voltage and a real voltage error is shown in (6).

$$\begin{bmatrix} V_{ref_a}^{*} \\ V_{ref_b}^{*} \\ V_{ref_c}^{*} \end{bmatrix} = V_{m_{max-peak}} \begin{bmatrix} \sin(\omega t) \\ \sin(\omega t - \frac{2\pi}{3}) \\ \sin(\omega t + \frac{2\pi}{3}) \end{bmatrix}$$
(4)

$$\begin{bmatrix} U_d \\ U_q \end{bmatrix} = \begin{bmatrix} \cos(\omega t) & \sin(\omega t) \\ -\sin(\omega t) & \cos(\omega t) \end{bmatrix} \begin{bmatrix} U_\alpha \\ U_\beta \end{bmatrix}$$
(5)

$$V_{SE_{dq}}^{*} = \left(V_{ref_{dq}}^{*} - V_{S_{dq}}\right) - \left(V_{L_{dq}} - V_{S_{dq}}\right) \tag{6}$$

Lastly, the equation (7) and (8) is used to transform from dq-domain to abc-domain of the compensation reference voltage,  $V_{SE,dq}^*$ . To produce appropriate gating pulses for the series VSC, the compensation reference voltage series compensator,  $V_{SE,abc}^*$  is using a pulse width modulation (PWM) as a voltage control algorithm method.

$$\begin{bmatrix} U_{\alpha} \\ U_{\beta} \end{bmatrix} = \begin{bmatrix} \sin(\omega t) & \cos(\omega t) \\ -\cos(\omega t) & \sin(\omega t) \end{bmatrix} \begin{bmatrix} U_{d} \\ U_{q} \end{bmatrix}$$
(7)

$$\begin{bmatrix} U_a \\ U_b \\ U_c \end{bmatrix} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & 0 \\ -\frac{1}{2} & \frac{\sqrt{3}}{2} \\ -\frac{1}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} U_\alpha \\ U_\beta \end{bmatrix}$$
(8)

## C. Shunt APF Compensator Controller

Next, the control scheme structure of the shunt APF compensator is based on the SRF technique implement the following sequence of transformation  $(abc-\alpha\beta0-dq0-\alpha\beta0-abc)$ displayed in Fig. 6. The suggested STF is used for load current harmonic extraction in the shunt APF compensator controller. Firstly, the three-phase load current is converted from *abc* domain to  $\alpha\beta0$ -domain using Clarke-matrix (1). In  $\alpha\beta0$ domain, it consists of a fundamental and harmonic element. Therefore, the fundamental (fund) element state as a  $i_{L,\alpha\beta}$  (fund) and the harmonic (har) element state as  $i_{L,\alpha\beta}$  (har) in  $\alpha\beta$ -domain. The extraction of a fundamental element,  $i_{L,\alpha\beta}$  (fund) is utilizing the STF transfer function shown in (2). Here,  $K_2$  indicates the constant gain parameter, and  $f_{c2}$  indicates the power system's cut-off frequency. In this study, the rating is determined 20 for  $K_2$  and 50 Hz for  $f_{c2}$ . The harmonic element,  $i_{L,\alpha\beta}(har)$  is obtained after removing the fundamental element from the distorted load current as the harmonic extraction algorithm method is shown in (9). Thus, the harmonic element is extracted indirectly. Secondly, the transformation in d-domain is expressed in (10) is used the synchronization phase reference of STF-UVG in term of  $sin(\omega t)$  and  $cos(\omega t)$  obtained in (3) and load current of a harmonic element in (9). For transformation in q-domain, the original load current and the synchronization phase reference is utilized shown in (11). Moreover, the DClink voltage regulation algorithm method is to stabilization the DC-link voltage capacitor obtained by calculating the appropriate  $i_{error,dc}$ . It is measured by minimizing the error acquired between total instantaneous DC-link voltage and reference DC-link voltage  $V_{dc,ref}$  with a PI controller. In this study, the value of proportional gain is set as 0.3, and integral gains are set as 2 of the PI controllers. Lastly, the equation (7) and (8) is used to transform from dq-domain to abc-domain of compensation reference current,  $i_{SH,abc}$ . The compensation reference voltage current is compared with measured compensation current in a hysteresis current controller algorithm method to produce appropriate gating pulses for the shunt VSC.

$$\begin{bmatrix} i_{L_{\alpha} (har)} \\ i_{L_{\beta} (har)} \end{bmatrix} = \begin{bmatrix} i_{L_{\alpha}} - i_{L_{\alpha} (fund)} \\ i_{L_{\beta}} - i_{L_{\beta} (fund)} \end{bmatrix}$$
(9)

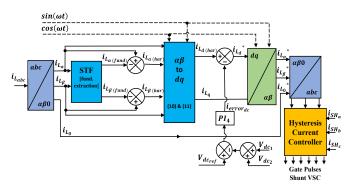


Fig. 6. Shunt APF Compensator control scheme structure

$$i_{L_{d}(har)} = i_{L_{\alpha}(har)} \sin(\omega t) - i_{L_{\beta}(har)} \cos(\omega t)$$
(10)

$$i_{L_q} = i_{L_\alpha} \cos(\omega t) + i_{L_\beta} \sin(\omega t)$$
(11)

## **III. SIMULATION RESULT**

The operating framework of the suggested UPQC is provided using the STF-SRF technique approach and evaluated using MATLAB Simulink (R2018a) technology. A common twolevel VSC for series and shunt APF is sharing common DC link capacitor of 9400 µF are utilized. For this study, the DC-link reference voltage is defined as 880 V. The output of the series and shunt compensator is then connected to the 5 mH L-typed filter, and the switching ripples are reduced. In the meantime, non-linear load imbalance is considered and consist of obtained distributed as three single phases in an unbalanced way. The capabilities of the suggested model are evaluated comparatively. The UPQC simulation using the proposed STF-SRF technique compared to the traditional SRF technique. Many case scenarios are performed in a steady-state condition where specific power quality issues of source voltage from PCC are considered under unbalanced and distorted source voltage conditions.

#### A. Balanced Voltage Sag

The capabilities of UPOC with STF-SRF technique is considered to deal with supply balanced voltage sag from PCC of MG for case scenario 1. Under this case scenario, all the obtained simulation waveform is displayed in Fig. 7-9. Furthermore, the highlighted comparative analysis for load voltage and source current between the suggested STF-SRF technique versus the traditional SRF technique is tabulated in Table I and II, respectively. It is observed that both techniques can precisely identify synchronization phase value  $\omega t$  in the shape of sawtooth waveform shape from balanced voltage sag of source voltage in Fig. 7. Thus, from Fig. 8(A), shown during 0.36s to 0.49s, source voltage from PCC has a voltage sag of 0.3 p.u from a standard magnitude of the source voltage. The mitigation process to keep the load voltage at the rated voltage by injected appropriate voltage in phase with the grid source voltage by series compensator is shown in Fig. 8(B). As the non-linear load is used, the distorted load current is shown in

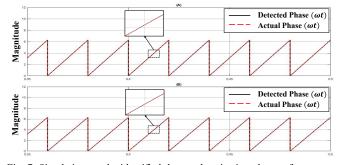


Fig. 7. Simulation results identified the synchronization phase reference,  $\omega t$  under balanced voltage sag and voltage swell, comparative with (A) suggested STF-*SRF* Technique (B) traditional SRF technique.

Fig. 8(D). The compensation of current reference by shunt compensator successfully mitigated harmonic current produced by non-linear load. The current supply has risen during voltage sag condition to retain power balance in the system but has regained sinusoidal as shown in Fig. 8(F). Thus, the THD of source current is measured between 0.4s to 0.45s when steadystate waveform condition. From Fig. 9 show the voltage across DC-link capacitor is reduced during 0.36s to 0.49s so that the load voltage can be supported at rated voltage during voltage sag condition. It also shows the UPQC is constructively regulated and retained DC-link voltage at the desired reference value of 880V (between the acceptance voltage range) after the source voltage's nominal condition. It shows the utilized PI controller for regulating the DC-link voltage has performed excellent and reliable with the suggested STF-SRF technique based on UPQC. Table I shows the suggested STF-SRF technique can minimize THD values of distorted load voltage that deal voltage sag from source voltage at PCC with optimum compensation magnitude and phase angle by the series compensator. In contrast, the traditional SRF technique

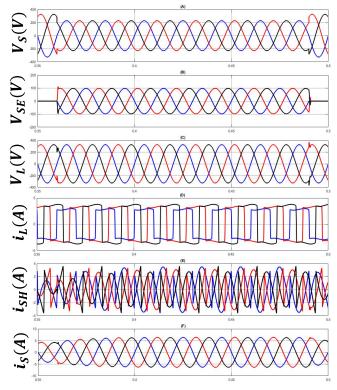


Fig. 8. Simulation waveform acquired under balanced voltage sag, with include (A) Three-phase source voltage (B) Compensated voltage (C) Load Voltage (D) Load Current (E) Compensated current (F) Source Current

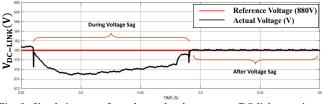


Fig. 9. Simulation waveform the total voltage across DC-link capacitor of UPQC acquired under balanced voltage sag.

compensation magnitude and phase angle do not effectively like the suggested technique. By referring to Table II, both techniques can reduce harmonic source current with THD rates fulfill with IEEE standard 519 (below than 5%). Furthermore, the vast phase differences between the source voltage and current can be reduced by both techniques. Lastly, both techniques lead to a proximity unity power factor of 0.999 can be obtained with a minimum phase difference.

## B. Balanced Voltage Swell

The capabilities of UPQC with STF-SRF technique is considered to deal with supply balanced voltage swell from PCC of MG for case scenario 2. Under this case scenario, all the obtained simulation waveform is displayed in Fig. 7 and Fig. 10-11. Furthermore, the highlighted comparative analysis for load voltage and source current between the STF-SRF and traditional SRF techniques is tabulated in Table I and II, respectively. It is observed that both techniques can precisely identify synchronization phase value  $\omega t$  in the shape of sawtooth waveform shape from balanced voltage swell of source voltage in Fig. 7. Thus, from Fig. 10(A), shown during 0.36s to 0.49s, the source voltage from PCC has a voltage swell of 0.3 p.u from a standard magnitude source voltage. The mitigation process to keep the load voltage at the rated voltage like in Fig. 10(C) by compensating appropriate voltage in phase with the grid voltage by series compensator shown in Fig. 10(B). As the non-linear load is used, that leads to the distorted load current. The injection of current reference by shunt

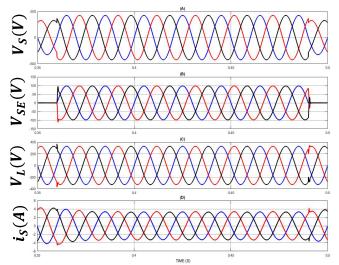


Fig. 10. Simulation waveform acquired under balanced voltage swell, with include (A) Three-phase source voltage (B) Compensated voltage (C) Load Voltage (D) Source Current

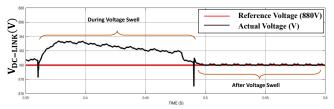


Fig. 11. Simulation waveform the total DC-link voltage of UPQC acquired under balanced voltage swell.

compensator successfully mitigated harmonic current produced by non-linear load. The current supply has decreased during voltage swell condition to retain power balance in the system but still has regained sinusoidal as shown in Fig. 10(D). Thus, the THD of source current is measured between 0.4s to 0.45s when steady-state waveform condition. Fig. 11 shows the voltage across the DC-link capacitor increases from 0.36s to 0.49s to maintain the load voltage at the rated voltage during voltage swell condition. It also shows the UPQC is constructively regulated and retained DC-link voltage at the desired reference value of 880V after the source voltage's nominal condition. It shows the utilized PI controller for regulating the DC-link voltage has performed excellent and reliable with the suggested STF-SRF technique based on UPQC. Table I shows the suggested STF-SRF technique can minimize THD values of distorted load voltage that deal voltage swell from source voltage at PCC with optimum compensation magnitude and phase angle by the series compensator. In contrast, the traditional SRF technique compensation magnitude and phase angle do not effectively like the suggested technique. By referring to Table II, both techniques can reduce harmonic source current with THD values complying with IEEE standard 519 (below than 5%). Furthermore, the large phase differences between the source current and voltage can be reduced by both techniques. Lastly, both techniques lead to an almost unity power factor of 0.999 can be obtained with a minimum phase difference.

# C. Balanced Voltage Harmonic

The capabilities of UPQC with STF-SRF technique is considered to deal with harmonic distorted supply voltage (a non-sinusoidal-balanced waveform) from PCC of MG for case scenario 3. Under this case scenario, all the obtained simulation waveform is displayed in Fig. 12-14. Furthermore, the highlighted comparative analysis for load voltage and source current between the suggested STF-SRF technique versus the traditional SRF technique is tabulated in Table I and II, respectively. From Fig. 12, it is observed that the suggested STF-SRF techniques can accurately detect synchronization phase value  $\omega t$  from balanced voltage harmonic of source voltage, where an identified phase value meets the desired phase value precisely. In comparison, there were inconsistencies between the identified and the desired phase

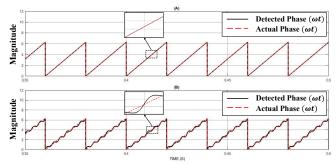


Fig. 12. Simulation result identified the synchronization reference phase,  $\omega t$  under balanced voltage harmonic, comparative with (A) suggested STF-*SRF* Technique (B) traditional SRF technique.

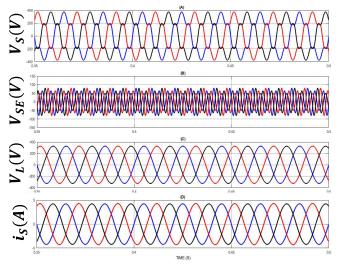


Fig. 13. Simulation waveform acquired under balanced voltage harmonic, with include (A) Three-phase source voltage (B) Compensated voltage (C) Load Voltage (D) Source Current.

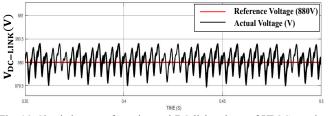


Fig. 14. Simulation waveform the total DC-link voltage of UPQC acquired under balanced voltage harmonic.

value for the traditional SRF technique. The identified phase value is observed oscillating with the resulting sawtooth waveform around the desired phase value. This means that where the source voltage suffers from distortion, the traditional SRF technique cannot function as required. From Fig. 13(A), shown during 0.35s to 0.5s can be observed that source voltage from PCC is distorted with harmonic (non-sinusoidal-balanced waveform). The mitigation process keeps the load voltage sinusoidal by compensating appropriate voltage in phase with grid voltage by series compensator shown in Fig. 13(B). As the non-linear load is used, that leads to the distorted load current. The injection of current reference by shunt compensator successfully mitigated harmonic current produced by nonlinear load. The current supply has regained sinusoidal can been noticed in Fig. 13(D). From Fig. 14 shows the UPQC is constructively regulated and retained DC-link capacitor voltage at the desired reference value of 880V by performing charging and discharging within the acceptance range of voltage during the compensation process. It shows the utilized PI controller for regulating the DC-link voltage has performed excellent and reliable with the suggested STF-SRF technique based on UPQC. Table I shows the suggested STF-SRF technique can minimize THD values of distorted load voltage that deal with harmonic issues from source voltage at PCC with optimum compensation magnitude and phase angle by the series compensator. In contrast, the traditional SRF technique can best get suitable compensation magnitude and phase angle but not effectively like the suggested technique. By referring to Table

II, the suggested STF-SRF technique can reduce harmonic supply current with THD values complying with IEEE standard 519 (below than 5%). Nevertheless, the traditional SRF technique fails to provide mitigation capabilities that reduce harmonic with THD values below that 5%. Furthermore, the large phase differences between the source current and voltage can be reduced by both techniques. When the suggested STF-SRF technique is utilized, an almost unity power factor of 0.999 can be obtained with a minimum phase difference. However, the traditional SRF technique can only secure a power factor up to 0.997. Some distortion is retained in the supply current after the mitigation process despite having a minimum phase difference.

## D. Unbalanced Voltage

The capabilities of UPQC with STF-SRF technique is considered to deal with supply unbalanced voltage from PCC of MG for case scenario 4. Under this case scenario, all the obtained simulation waveform is displayed in Fig. 15-17. Furthermore, the highlighted comparative analysis for load voltage and source current between the suggested STF-SRF technique versus the traditional SRF technique is tabulated in Table I and II, respectively. From Fig. 15, it is observed that the suggested STF-SRF techniques can accurately detect synchronization phase value  $\omega t$  from the unbalanced voltage of source voltage, where an identified phase value meets the desired phase value precisely. In comparison, there were slight inconsistencies and small difference between the identified and the desired phase value for the traditional SRF technique. Thus, the traditional SRF technique cannot perform effectively as desired when dealing with unbalance three-phase voltage magnitude of the source voltage. From Fig. 16(A), shown during 0.35s to 0.5s can be observed that source voltage from PCC having an unbalanced magnitude of the voltage across a three-phase supply. The mitigation process keeps the load voltage at rated voltage with balanced magnitude across threephase by compensated appropriate voltage in phase with the grid voltage by series compensator shown in Fig. 16(B). As the non-linear load is used, that leads to the distorted load current. The injection of current reference by shunt compensator successfully mitigated harmonic current produced by nonlinear load. The current supply has regained sinusoidal can been noticed in Fig. 16(D). From Fig. 17 shows the UPQC is constructively regulated and retained DC-link capacitor voltage

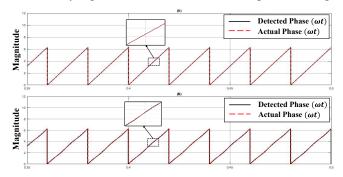


Fig. 15. Simulation result identified the synchronization reference phase,  $\omega t$  under unbalanced voltage, comparative with (A) suggested STF-SRF Technique (B) traditional SRF technique.

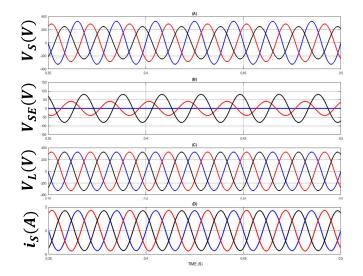


Fig. 16. Simulation waveform acquired under Unbalanced Voltage, with include (A) Three-phase source voltage (B) Compensated voltage (C) Load Voltage (D) Source Current.

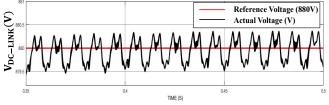


Fig. 17. Simulation waveform the total DC-link voltage of UPQC acquired under unbalanced voltage.

at the desired reference value of 880V by performing charging and discharging within the acceptance range of voltage during the compensation process. It shows the utilized PI controller for regulating the DC-link voltage has performed excellent and reliable with the suggested STF-SRF technique based on UPQC. Table I shows the suggested STF-SRF technique can minimize THD values of distorted load voltage that deal unbalanced magnitude issues across three-phase from source voltage at PCC with optimum compensation magnitude and phase angle by the series compensator. However, the traditional SRF technique compensation magnitude and phase angle do not effectively like the suggested technique. By referring to Table II, the suggested STF-SRF technique can reduce harmonic supply current with THD values complying with IEEE standard 519 (below than 5%). However, the traditional SRF technique fails to provide mitigation capabilities that reduce harmonic with THD values below that 5%. Furthermore, the large phase differences between the source current and voltage can be reduced by both techniques. When the suggested STF-SRF technique is utilized, an almost unity power factor of 0.999 can be obtained with a minimum phase difference. However, the traditional SRF technique can only secure a power factor up to 0.998 because some distortion retained in the supply current after the mitigation process despite having a minimum phase difference.

## E. Non-sinusoidal-Unbalanced Voltage

The capabilities of UPQC with STF-SRF technique is

considered to deal with supply non-sinusoidal (harmonic distorted) and unbalanced magnitude voltage across threephase from PCC of MG for case scenario 5. Under this case scenario, all the obtained simulation waveform is displayed in Fig. 18-20. Furthermore, the highlighted of comparative analysis for load voltage and source current between suggested STF-SRF technique versus traditional SRF technique is tabulated in Table I and II, respectively. From Fig. 18 observed that the suggested STF-SRF techniques could accurately detect synchronization phase value  $\omega t$  from non-sinusoidalunbalanced voltage of source voltage. An identified phase value meets the desired phase value precisely. In comparison, there were slight inconsistencies between the identified and the desired phase value for the traditional SRF technique. Thus, the traditional SRF technique cannot perform effectively when dealing with unbalance three-phase voltage magnitude and harmonic distorted of source voltage. From Fig. 19(A), shown during 0.35s to 0.5s can be observed that source voltage from PCC having an unbalanced magnitude of the voltage across three-phase supply with distorted harmonic. The mitigation process to keep the load voltage at rated voltage with balanced magnitude across three-phase by compensated appropriate voltage in phase with the grid voltage by series compensator shown in Fig. 19(B). As the non-linear load is used, that leads to the distorted load current. The injection of current reference by shunt compensator successfully mitigated harmonic current produced by non-linear load. The source current has regained sinusoidal can been noticed in Fig. 19(D). Furthermore, from Fig. 20 shows the UPQC is constructively regulated and retained DC-link capacitor voltage at the desired reference value of 880V by performing charging and discharging within the acceptance range of voltage during the compensation process. It is showing the utilize PI controller for regulating the DC-link voltage have performed excellent and reliable with suggested STF-SRF technique based on UPQC. Table I shows the suggested STF-SRF technique can minimize THD values of distorted load voltage that deal unbalanced magnitude issues across three-phase with distorted harmonic from supply voltage at PCC with optimum compensation magnitude and phase angle by the series compensator. However, the traditional SRF technique compensation magnitude and phase angle do not effectively like suggested technique. By referring to Table II, the suggested STF-SRF technique can reduce harmonic supply

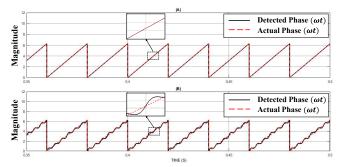


Fig. 18. Simulation result identified the synchronization reference phase,  $\omega t$  under non-sinusoidal-unbalanced voltage, comparative with (A) suggested STF-*SRF* Technique (B) traditional SRF technique.

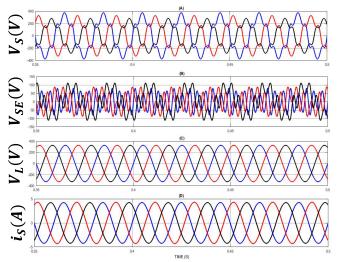


Fig. 19. Simulation waveform acquired under non-sinusoidal-unbalanced voltage, with include (A) Three-phase source voltage (B) Compensated voltage (C) Load Voltage (D) Source Current.

current with THD values complying with IEEE standard 519 (below than 5%). However, the traditional SRF technique fails to provide mitigation capabilities that reduce harmonic with THD values below that 5%. Furthermore, the large phase differences that emerged between the source current and voltage can reduce by both techniques. When the suggested STF-SRF technique is utilized, almost unity power factor of 0.999 can be obtained with a minimum phase difference. However, the traditional SRF technique can only secure a power factor up to 0.998 because some distortion retained in the

supply current after the mitigation process despite having a minimum phase difference.

Generally based on case scenario 1 to 5, all the result shows the suggested STF-SRF technique design method based on UPQC and working performance can be accepted especially during dealing with unbalance and distorted source voltage condition. Based on Table I and II, the suggested technique is revealed the ideal sinusoidal load voltage has been regained with low THD values below than 0.29% with appropriate compensation magnitude. Other than that, by complying the IEEE standard 519, the most distorted source current has been recovered with low THD values between 1.99% to 3.15% with minimizing the phase differences to achieving 0.999 of power factor which is closed to unity. In addition, all DC-link capacitor voltages have been controlled as required, meaning the UPQC can properly mitigate power quality issues in the

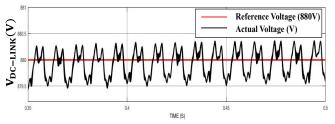


Fig. 20. Simulation waveform the total DC-link voltage of UPQC acquired under non-sinusoidal-unbalanced voltage.

operating power system.

Col	MPARATIV	E LOAD V	OLTAGE AN	ALYSIS C	F THE SUC	GESTED	UPQC us	ING STF-	SRF TECH	INIQUE AI	ND TRADIT	TIONAL SI	-	· ·	
	Balance Voltage Sag			Balance Voltage Swell			Balance Voltage			Unbalance Voltage			Non-sinusoidal-		
Analysis							Harmonic						Unbalanced Voltage		
Parameter	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase
	а	b	С	а	b	С	а	b	С	а	b	С	а	b	С
						Before (	Connectin	g UPQC							
THD value (%)	17.65	19.36	19.28	16.84	17.63	17.03	22.62	24.11	24.22	12.14	13.54	13.33	24.63	26.31	25.74
Phase Angle (°)	0	-120	120	0	-120	120	0	-120	120	0	-120	120	0	-120	120
				After Co	onnecting	UPQC w	ith the su	ggested S	ΓF-SRF to	echnique					
THD value (%)	0.25	0.25	0.25	0.25	0.25	0.25	0.28	0.28	0.28	0.29	0.29	0.29	0.29	0.29	0.29
Phase Angle (°)	0	-120	120	0	-120	120	0	-120	120	0	-120	120	0	-120	120
Compensation Magnitude (%)	100	99.64	99.814	100	100.3	100.4	100	100.1	99.78	100	99.89	99.81	100.1	99.78	99.64
				After	Connectir	ng UPQC	with the	traditional	SRF tecl	nnique					
THD value (%)	0.72	1.86	1.98	0.61	1.54	2.00	0.31	0.31	0.32	0.65	1.03	1.11	0.93	2.33	2.51
Phase Angle (°)	0	-120.2	119.9	0	-120.2	199.9	0	-120.2	199.9	0	-120.2	199.9	0	-120.2	199.9
Compensation Magnitude (%)	100	99.2	99.43	100	100.6	100.4	99.97	99.48	100.2	99.94	99.68	100.4	99.70	99.48	100.5

TABLE I

	Balar	nce Voltag	e Sag	NALYSIS OF THE SUGGESTED Balance Voltage Swell			Balance Voltage			Unba	alance Vo	ltage	Non-sinusoidal-		
Analysis		2			0		Harmonic			8			Unbalanced Voltage		
Parameter	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase	Phase
	а	b	С	а	b	с	а	b	С	а	b	С	а	b	С
Before Connecting UPQC															
THD value (%)	48.49	46.57	49.43	48.49	46.57	49.43	53.28	52.92	53.28	48.49	46.57	49.43	59.49	57.57	59.43
Phase Difference (°)	10.6	9.1	9.8	10.6	9.1	9.8	12.5	11.3	9.2	11.6	9.4	10.3	10.2	9.8	9.3.7
Power Factor	0.913	0.860	0.920	0.913	0.860	0.920	0.808	0.812	0.905	0.903	0.810	0.910	0.906	0.850	0.926
After Connecting UPQC with the suggested STF-SRF technique															
THD value (%)	1.99	2.13	2.76	1.98	2.59	2.88	2.09	2.67	3.15	2.79	2.86	3.01	2.81	2.69	2.99
Phase Difference (°)	0.3	0.2	0.4	0.2	0.2	0.5	0.4	0.3	0.5	0.2	0.3	0.3	0.2	0.4	0.3
Power Factor	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999	0.999
After Connecting UPQC with the traditional SRF technique															
THD value (%)	2.89	2.80	3.00	2.66	2.50	2.81	4.12	5.75	6.14	5.09	5.11	5.69	5.87	5.91	6.04
Phase Difference (°)	0.5	0.3	0.6	0.4	0.4	0.7	0.5	0.4	0.7	0.3	0.5	0.4	0.5	0.6	0.4
Power Factor	0.999	0.999	0.999	0.999	0.999	0.999	0.997	0.997	0.996	0.998	0.996	0.997	0.997	0.996	0.995

#### TABLE II CURRENT ANALYSIS OF THE SUGGESTED UPQC USING STF-SRF TECHNIQUE AND TRADITIONAL SRF TECHNIQUE:

## IV. CONCLUSION

The use of the STF-SRF technique-based on UPQC improves standard UPQC in dealing with adverse source voltage conditions. Besides that, this technique is applied effectively in the UPQC controller scheme. The modified SRF technique integrated with STF-UVG as phase synchronization which can lead to UPQC controller generated reference current and voltage in phase with the operating power system. Furthermore, the UPQC with STF-SRF technique is formed without relying on the complexity of PLL components. The compensation of current and voltage is performed successfully following the adverse source voltage condition and consideration of nonlinear load. This is to ensure the system stability and obtains almost unity power factor. It is verified that the harmonics are minimized in source current by following the IEEE-519 standard. On top of that, the ideal sinusoidal load voltage has been regained with low THD values with appropriate compensation magnitude. Therefore, the suggested STF-SRF technique based on the UPQC system can enhance the overall performance and more reliable to mitigate power quality issues by considering dealing with adverse source voltage conditions in the power grid system.

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